

Discrete-Time H_∞ Control Synthesis

Lecture #3

Hisaya Fujioka

fujioka@{i.kyoto-u.ac.jp, math.kth.se}

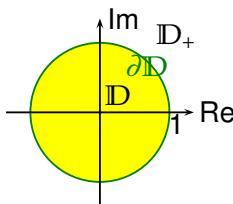
Graduate School of Informatics, Kyoto University

Discrete-Time H_∞ Control Synthesis – p.1/27

H_∞ Norm (1/2)

Notation

- $\mathbb{D} := \{ z \mid z \in \mathbb{C}, |z| < 1 \}$
- $\partial\mathbb{D} := \{ z \mid z \in \mathbb{C}, |z| = 1 \}$
- $\mathbb{D}_+ := \{ z \mid z \in \mathbb{C}, |z| > 1 \}$



Definition

$$(i) \quad \mathbf{H}_\infty(\mathbb{D}_+) := \left\{ \hat{T} \mid \sup_{z \in \mathbb{D}_+} \sigma_{\max}(\hat{T}[z]) < \infty \right\}$$

(ii) Given $\hat{T} \in \mathbf{H}_\infty$,

$$\|\hat{T}\|_\infty := \sup_{z \in \mathbb{D}_+} \sigma_{\max}(\hat{T}[z])$$

Discrete-Time H_∞ Control Synthesis – p.3/27

Outline

- Problem Formulation
 - H_∞ Norm
 - H_∞ Control Synthesis
- Reduction to Continuous-Time Problem
 - Bilinear Transformation
- LMI Solution
 - Bounded Real Lemma
 - Elimination Lemma
 - LMI Feasibility Condition
 - Controller Construction

Discrete-Time H_∞ Control Synthesis – p.2/27

H_∞ Norm (2/2)

Property

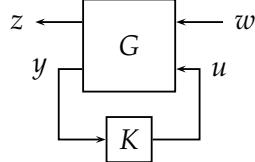
$$\hat{T}[z] = C(zI - A)^{-1}B + D$$

- (i) $\text{eig}(A) \subset \mathbb{D} \Rightarrow \hat{T} \in \mathbf{H}_\infty$
- (ii) $\hat{T} \in \mathbf{H}_\infty \Rightarrow \text{eig}(A) \subset \mathbb{D}$ if stabilizable and detectable
- (iii) Given $T \in \mathbf{H}_\infty$,

$$\begin{aligned} \|\hat{T}\|_\infty &= \sup_{z \in \partial\mathbb{D}} \sigma_{\max}(\hat{T}[z]) \\ &= \sup_{\theta \in [0, 2\pi]} \sigma_{\max}(\hat{T}[e^{j\theta}]) \end{aligned}$$

Discrete-Time H_∞ Control Synthesis – p.4/27

Design Framework



■ Given: Generalized plant G

$$\hat{G}[z] = \begin{bmatrix} C_1 \\ C_2 \end{bmatrix} (zI - A)^{-1} \begin{bmatrix} B_1 & B_2 \end{bmatrix} + \begin{bmatrix} D_{11} & D_{12} \\ D_{21} & 0 \end{bmatrix}$$

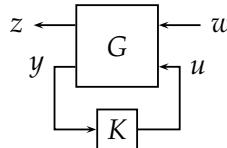
■ Find: Controller K

$$\hat{K}[z] = C_K(zI - A_K)^{-1}B_K + D_K$$

■ $z[z] = \mathcal{F}_\ell(\hat{G}, \hat{K})[z] w[z]$

Discrete-Time H_∞ Control Synthesis – p.5/27

H_∞ Control Synthesis



$$\mathcal{F}_\ell(\hat{G}, \hat{K})[z] = C_{cl}(zI - A_{cl})^{-1}B_{cl} + D_{cl}$$

$$\left[\begin{array}{cc|c} A_{cl} & B_{cl} & B_1 \\ 0 & 0 & 0 \\ \hline C_{cl} & D_{cl} & D_{11} \end{array} \right] := \left[\begin{array}{cc|c} A & 0 & B_1 \\ 0 & 0 & 0 \\ \hline C_1 & 0 & D_{11} \end{array} \right] + \left[\begin{array}{cc|c} B_2 & 0 & 0 \\ 0 & I & 1 \\ \hline D_{12} & 0 & 0 \end{array} \right] \left[\begin{array}{cc|c} D_K & C_K & 0 \\ B_K & A_K & I \\ \hline 0 & 0 & 0 \end{array} \right] \left[\begin{array}{cc|c} C_2 & 0 & D_{21} \\ 0 & I & 0 \end{array} \right]$$

H_∞ control synthesis

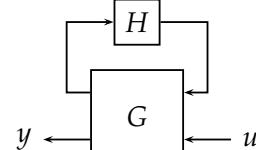
Given G . Find K such that

■ $\text{eig}(A_{cl}) \subset \mathbb{D}$

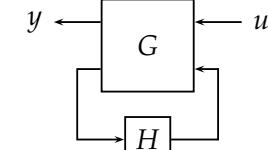
■ $\|\mathcal{F}_\ell(\hat{G}, \hat{K})\|_\infty < 1$

Discrete-Time H_∞ Control Synthesis – p.7/27

LFT (Linear Fractional Transformation)



$\mathcal{F}_u(G, H)$: upper LFT



$\mathcal{F}_\ell(G, H)$: lower LFT

$$\mathcal{F}_u(G, H) := G_{22} + G_{21}H(I - G_{11}H)^{-1}G_{12}$$

$$\mathcal{F}_\ell(G, H) := G_{11} + G_{12}H(I - G_{22}H)^{-1}G_{21}$$

Example

$$C(zI - A)^{-1}B + D = \mathcal{F}_u(M, z^{-1}I), \quad M := \begin{bmatrix} A & B \\ C & D \end{bmatrix}$$

Discrete-Time H_∞ Control Synthesis – p.6/27

Outline

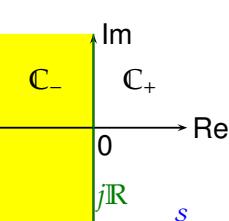
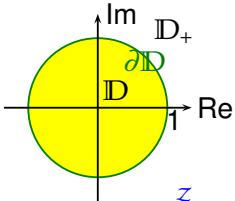
- Problem Formulation
 - H_∞ Norm
 - H_∞ Control Synthesis
- Reduction to Continuous-Time Problem
 - Bilinear Transformation
- LMI Solution
 - Bounded Real Lemma
 - Elimination Lemma
 - LMI Feasibility Condition
 - Controller Construction

Discrete-Time H_∞ Control Synthesis – p.8/27

Bilinear Transformation (1/2)

Bilinear Transformation

$$z = \frac{1+s}{1-s}$$



\mathcal{B}

Property: Given \hat{G} . Let $\hat{G}_c(s) := \hat{G} \left[\frac{1+s}{1-s} \right]$

- $\hat{G} \in \mathbf{H}_{\infty}(\mathbb{D}_+) \Leftrightarrow \hat{G}_c \in \mathbf{H}_{\infty}(\mathbb{C}_+)$
- Suppose $\hat{G} \in \mathbf{H}_{\infty}(\mathbb{D}_+)$. $\|\hat{G}\|_{\infty} = \|\hat{G}_c\|_{\infty}$

Discrete-Time \mathbf{H}_{∞} Control Synthesis – p.9/27

Bilinear Transformation (2/2)

Given $\hat{G}[z] = C(zI - A)^{-1}B + D$ with $-1 \notin \text{eig}(A)$

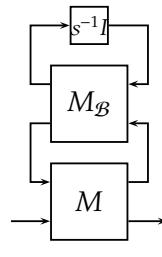
Fact: $\hat{G}_c(s) := \hat{G} \left[\frac{1+s}{1-s} \right] = C_c(sI - A_c)^{-1}B_c + D_c$

$$\begin{bmatrix} A_c & B_c \\ C_c & D_c \end{bmatrix} := \begin{bmatrix} (A - I)(A + I)^{-1} & \sqrt{2}(A + I)^{-1}B \\ \sqrt{2}C(A + I)^{-1} & \hat{G}[-1] \end{bmatrix}$$

Proof: Notice

- $\hat{G}[z] = \mathcal{F}_u(M, z^{-1}I)$, $M := \begin{bmatrix} A & B \\ C & D \end{bmatrix}$
- $z^{-1}I = \frac{1-s}{1+s}I = \mathcal{F}_u(M_B, s^{-1}I)$, $M_B := \begin{bmatrix} -I & \sqrt{2}I \\ \sqrt{2}I & -I \end{bmatrix}$,

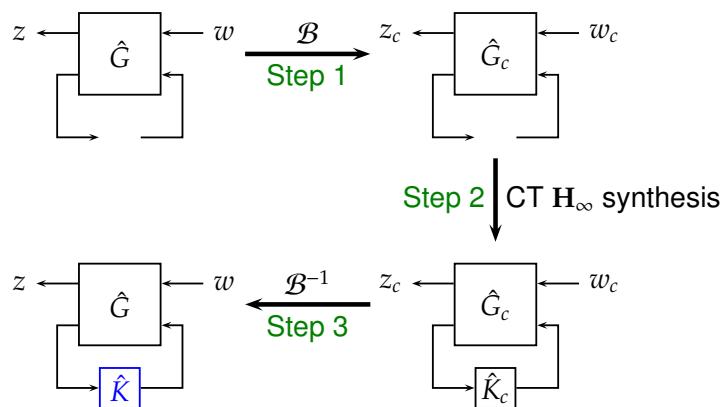
$$\hat{G}_c(s) = \mathcal{F}_u(M, \mathcal{F}_u(M_B, s^{-1}I)) = \mathcal{F}_u(M_B \star M, s^{-1}I)$$



Discrete-Time \mathbf{H}_{∞} Control Synthesis – p.10/27

Reduction to Continuous-Time \mathbf{H}_{∞} Control Synthesis

Design Procedure



Discrete-Time \mathbf{H}_{∞} Control Synthesis – p.11/27

Outline

- Problem Formulation
 - \mathbf{H}_{∞} Norm
 - \mathbf{H}_{∞} Control Synthesis
- Reduction to Continuous-Time Problem
 - Bilinear Transformation
- LMI Solution
 - Bounded Real Lemma
 - Elimination Lemma
 - LMI Feasibility Condition
 - Controller Construction

Discrete-Time \mathbf{H}_{∞} Control Synthesis – p.12/27

Bounded Real Lemma

Notation

- $A > 0$: $A \in \mathbb{R}^{n \times n}$ is positive definite
- $A > B$: $A - B > 0$

Lemma Given $\hat{T}[z] = C(zI - A)^{-1}B + D$, $\gamma > 0$

(i) $\text{eig}(A) \subset \mathbb{D}$ and $\|\hat{T}\|_{\infty} < \gamma$

\Leftrightarrow (ii) $\exists P = P^T > 0$,

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix}^T \begin{bmatrix} P & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A & B \\ C & D \end{bmatrix} - \begin{bmatrix} P & 0 \\ 0 & \gamma^2 I \end{bmatrix} < 0$$

- Analysis via convex optimization

Discrete-Time H_{∞} Control Synthesis – p.13/27

Elimination Lemma

Notation: Given $A \in \mathbb{R}^{n \times m}$, $\text{rank}(A) = r$

$A^\perp \in \mathbb{R}^{(n-r) \times n}$ is any matrix satisfying

$$A^\perp A = 0, \quad A^\perp A^{\perp T} > 0$$

Lemma Given $L \in \mathbb{R}^{n \times m}$, $R \in \mathbb{R}^{p \times n}$, and $M = M^T \in \mathbb{R}^{n \times n}$

\Leftrightarrow (i) $\exists Z \in \mathbb{R}^{m \times p}$, $LZR + (LZR)^T + M < 0$

$$\left\{ \begin{array}{l} L^\perp M L^{\perp T} < 0 \quad \text{or} \quad LL^T > 0 \\ R^{\perp T} M R^{\perp T} < 0 \quad \text{or} \quad R^T R > 0 \end{array} \right.$$

Discrete-Time H_{∞} Control Synthesis – p.15/27

Schur Complement

Lemma SAE

- $\begin{bmatrix} M_1 & M_3 \\ M_3^T & M_2 \end{bmatrix} > 0$
- $M_1 > 0$ and $M_2 - M_3^T M_2^{-1} M_3 > 0$
- $M_2 > 0$ and $M_1 - M_3 M_2^{-1} M_3^T > 0$

Proof: easy

Discrete-Time H_{∞} Control Synthesis – p.14/27

LMI Feasibility Condition

Theorem Given G . SAE

(i) \exists a solution to H_{∞} control synthesis problem

(ii) $\exists X = X^T > 0$, $Y = Y^T > 0$,

$$\left\{ \begin{array}{l} \mathcal{B}^\perp \left(\mathcal{A} \begin{bmatrix} X & 0 \\ 0 & I \end{bmatrix} \mathcal{A}^T - \begin{bmatrix} X & 0 \\ 0 & I \end{bmatrix} \right) \mathcal{B}^{\perp T} < 0 \\ \mathcal{C}^{\perp T} \left(\mathcal{A}^T \begin{bmatrix} Y & 0 \\ 0 & I \end{bmatrix} \mathcal{A} - \begin{bmatrix} Y & 0 \\ 0 & I \end{bmatrix} \right) \mathcal{C}^{\perp} < 0 \\ \begin{bmatrix} X & I \\ I & Y \end{bmatrix} \geq 0 \end{array} \right.$$

where $\mathcal{A} := \begin{bmatrix} A & B_1 \\ C_1 & D_{11} \end{bmatrix}$, $\mathcal{B} := \begin{bmatrix} B_2 \\ D_{12} \end{bmatrix}$, $\mathcal{C} := \begin{bmatrix} C_2 & D_{21} \end{bmatrix}$

Discrete-Time H_{∞} Control Synthesis – p.16/27

Sketch of Proof (1/3)

- Step 1: Bounded Real Lemma

$$\begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix}^T \begin{bmatrix} P_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} - \begin{bmatrix} P_{cl} & 0 \\ 0 & I \end{bmatrix} < 0$$

$$\begin{bmatrix} \begin{bmatrix} P_{cl} & 0 \\ 0 & I \end{bmatrix} & \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} \\ \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} & \begin{bmatrix} P_{cl}^{-1} & 0 \\ 0 & I \end{bmatrix} \end{bmatrix}^T > 0$$

- Step 2: Notice that

$$\begin{bmatrix} I & 0 & 0 \\ 0 & 0 & I \\ 0 & I & 0 \end{bmatrix} \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ 0 & 0 & I \\ 0 & I & 0 \end{bmatrix} = \begin{bmatrix} \mathcal{A} & 0 \\ 0 & 0 \end{bmatrix} + \begin{bmatrix} \mathcal{B} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} D_K & C_K \\ B_K & A_K \end{bmatrix} \begin{bmatrix} C & 0 \\ 0 & I \end{bmatrix}$$

Discrete-Time H_∞ Control Synthesis – p.17/27

Sketch of Proof (3/3)

$$\mathcal{L}\mathbf{Z}_K\mathcal{R} + (\mathcal{L}\mathbf{Z}_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$$

$$\mathcal{L} := \begin{bmatrix} 0 \\ \begin{bmatrix} \mathcal{B} & 0 \\ 0 & I \end{bmatrix} \end{bmatrix}, \quad \mathcal{R} := \begin{bmatrix} \begin{bmatrix} C & 0 \\ 0 & I \end{bmatrix} & 0 \end{bmatrix}, \quad \mathcal{M}(\mathcal{P}_{cl}) := \begin{bmatrix} \begin{array}{c|c} Y & 0 \\ \hline 0 & I \\ \hline P_3^T & 0 \end{array} & \begin{bmatrix} \mathcal{A}^T & 0 \\ 0 & 0 \end{bmatrix} \\ \begin{bmatrix} \mathcal{A} & 0 \\ 0 & 0 \end{bmatrix} & \begin{array}{c|c} X & 0 \\ \hline 0 & I \\ \hline Q_3^T & 0 \end{array} \end{bmatrix}$$

- Step 3: Elimination Lemma for \mathbf{Z}_K with

$$\mathcal{L}^\perp = \begin{bmatrix} I & 0 \\ 0 & [\mathcal{B}^\perp & 0] \end{bmatrix}, \quad \mathcal{R}^{T\perp} = \begin{bmatrix} [\mathcal{C}^{T\perp} & 0] & 0 \\ 0 & I \end{bmatrix}$$

Discrete-Time H_∞ Control Synthesis – p.19/27

Sketch of Proof (2/3)

$$\begin{bmatrix} \begin{array}{c|c} Y & 0 \\ \hline 0 & I \\ \hline P_3^T & 0 \end{array} & \begin{bmatrix} P_3 \\ 0 \\ 0 \end{bmatrix} \\ \begin{bmatrix} \mathcal{A} & 0 \\ 0 & 0 \end{bmatrix} + \begin{bmatrix} \mathcal{B} & 0 \\ 0 & I \end{bmatrix} \mathbf{Z}_K \begin{bmatrix} C & 0 \\ 0 & I \end{bmatrix} & \begin{array}{c|c} C^T & 0 \\ \hline 0 & I \\ \hline 0 & 0 \end{array} \end{bmatrix} \begin{bmatrix} \mathcal{B}^T & 0 \\ 0 & I \end{bmatrix} \Bigg) > 0$$

$$\text{where } \mathcal{P}_{cl} := \begin{bmatrix} Y & P_3 \\ P_3^T & P_2 \end{bmatrix}, \quad \mathcal{P}_{cl}^{-1} := \begin{bmatrix} X & Q_3 \\ Q_3^T & Q_2 \end{bmatrix}, \quad \mathbf{Z}_K := \begin{bmatrix} D_K & C_K \\ B_K & A_K \end{bmatrix}$$

$$\begin{aligned} Y &= (X - Q_3 Q_2^{-1} Q_3^T)^{-1} \\ \Rightarrow X - Y^{-1} &= Q_3 Q_2^{-1} Q_3^T \geq 0 \\ \Rightarrow \begin{bmatrix} X & I \\ I & Y \end{bmatrix} &\geq 0 \end{aligned}$$

Discrete-Time H_∞ Control Synthesis – p.18/27

Outline

- Problem Formulation
 - H_∞ Norm
 - H_∞ Control Synthesis
- Reduction to Continuous-Time Problem
 - Bilinear Transformation
- LMI Solution
 - Bounded Real Lemma
 - Elimination Lemma
 - LMI Feasibility Condition
 - Controller Construction

Discrete-Time H_∞ Control Synthesis – p.20/27

Recovering P_{cl}

- Solution to synthesis LMI (X, Y) satisfying $\begin{bmatrix} X & I \\ I & Y \end{bmatrix} > 0$

- otherwise replace by $(X + \varepsilon I, Y + \varepsilon I)$
- $(X + \varepsilon I, Y + \varepsilon I)$ solves synthesis LMI for small $\varepsilon > 0$

- Fact: $\exists P_2, Q_2$ such that

$$\begin{bmatrix} Y & N \\ N^T & P_2 \end{bmatrix} > 0, \quad \begin{bmatrix} X & M \\ M^T & Q_2 \end{bmatrix} > 0, \quad \begin{bmatrix} Y & N \\ N^T & P_2 \end{bmatrix} \begin{bmatrix} X & M \\ M^T & Q_2 \end{bmatrix} = I$$

where $MN^T := I - XY$ P_{cl} P_{cl}^{-1}

Note: M, N are

- nonsingular
- computable via SVD

Discrete-Time H_∞ Control Synthesis – p.21/27

Controller Construction

- Naive Algorithm:

- Substitute P_{cl} back to $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$
- Solve as an LMI in Z_K (feasible !)
- Works well, but numerically expensive

- Sophisticated algorithm:

- Without recovering P_{cl} explicitly
- Decomposing $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$ by using

$$P_{cl} \begin{bmatrix} X & I \\ M^T & 0 \end{bmatrix} = \begin{bmatrix} I & Y \\ 0 & N^T \end{bmatrix}$$

$$\begin{bmatrix} X & I \\ M^T & 0 \end{bmatrix}^T P_{cl} \begin{bmatrix} X & I \\ M^T & 0 \end{bmatrix} = \begin{bmatrix} X & I \\ I & Y \end{bmatrix}$$

$$\begin{bmatrix} I & Y \\ 0 & N^T \end{bmatrix}^T P_{cl}^{-1} \begin{bmatrix} I & Y \\ 0 & N^T \end{bmatrix} = \begin{bmatrix} X & I \\ I & Y \end{bmatrix}$$

Discrete-Time H_∞ Control Synthesis – p.22/27

Decomposing $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$ (1/3)

Lemma SAE

(i) $\exists Z_K$ such that $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$

(ii) $\exists (\tilde{B}_K, \tilde{C}_K, D_K)$ such that

$$\begin{bmatrix} X & L_X(\tilde{C}_K) \\ L_X^T(\tilde{C}_K) & \Delta(D_K) \end{bmatrix} > 0, \quad \begin{bmatrix} Y & L_Y(\tilde{B}_K) \\ L_Y^T(\tilde{B}_K) & \Delta(D_K) \end{bmatrix} > 0$$

$$\text{where } \begin{bmatrix} L_X(\tilde{C}_K) \\ L_Y(\tilde{B}_K) \\ \Delta(D_K) \end{bmatrix} := \begin{bmatrix} \begin{bmatrix} I & 0 \\ \tilde{B}_K C + \begin{bmatrix} Y & 0 \end{bmatrix} \mathcal{A} & \tilde{C}_K^T \mathcal{B}^T + \begin{bmatrix} X & 0 \end{bmatrix} \mathcal{A}^T \\ \begin{bmatrix} Y & 0 \\ 0 & I \end{bmatrix} & (A + B D_K C)^T \\ \mathcal{A} + B D_K C & \begin{bmatrix} X & 0 \\ 0 & I \end{bmatrix} \end{bmatrix}$$

Discrete-Time H_∞ Control Synthesis – p.23/27

Decomposing $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^T + \mathcal{M}(\mathcal{P}_{cl}) > 0$ (2/3)

Sketch of Proof:

(i) $\Leftrightarrow \exists (\tilde{A}_K, \tilde{B}_K, \tilde{C}_K, D_K)$ such that

$$\begin{bmatrix} \begin{bmatrix} X & \tilde{A}_K^T \\ \tilde{A}_K & Y \end{bmatrix} & \begin{bmatrix} L_X(\tilde{C}_K) \\ L_Y(\tilde{B}_K) \end{bmatrix} \\ \begin{bmatrix} L_X^T(\tilde{C}_K) & L_Y^T(\tilde{B}_K) \end{bmatrix} & \Delta(D_K) \end{bmatrix} > 0$$

by taking

$$\begin{bmatrix} \tilde{A}_K & \tilde{B}_K \\ \tilde{C}_K & D_K \end{bmatrix} = \begin{bmatrix} Y & N & 0 \\ 0 & 0 & I \end{bmatrix} \left[\begin{bmatrix} A & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} + \begin{bmatrix} B_2 & 0 \\ 0 & I \end{bmatrix} Z_K \begin{bmatrix} C_2 & 0 & I \\ 0 & I & 0 \end{bmatrix} \right] \begin{bmatrix} X & 0 \\ M^T & 0 \\ 0 & I \end{bmatrix}$$

\Leftrightarrow (ii): Set $\tilde{A}_K = L_Y(\tilde{B}_K) \Delta^{-1}(D_K) L_X^T(\tilde{C}_K)$

Discrete-Time H_∞ Control Synthesis – p.24/27

Decomposing $\mathcal{L}Z_K\mathcal{R} + (\mathcal{L}Z_K\mathcal{R})^\top + \mathcal{M}(\mathcal{P}_{cl}) > 0$ (3/3)

Given solution (X, Y) to synthesis LMI

Lemma $\exists D_K$ such that $\Delta(D_K) > 0$

- $\exists \tilde{B}_K$ such that $\begin{bmatrix} Y & L_Y(\tilde{B}_K) \\ L_Y^\top(\tilde{B}_K) & \Delta(D_K) \end{bmatrix} > 0$

- $\exists \tilde{C}_K$ such that $\begin{bmatrix} X & L_X(\tilde{C}_K) \\ L_X^\top(\tilde{C}_K) & \Delta(D_K) \end{bmatrix} > 0$

Sketch of proof: Invoking Elimination Lemma for \tilde{B}_K ,

$$\begin{bmatrix} Y & L_Y(\tilde{B}_K) \\ L_Y^\top(\tilde{B}_K) & \Delta(D_K) \end{bmatrix} > 0 \Leftrightarrow \left\{ \begin{array}{l} \Delta(D_K) > 0 \\ C^{\top\perp} \left(\mathcal{A}^\top \begin{bmatrix} Y & 0 \\ 0 & I \end{bmatrix} \mathcal{A} - \begin{bmatrix} Y & 0 \\ 0 & I \end{bmatrix} \right) C^{\top\perp\top} < 0 \end{array} \right.$$

Discrete-Time H_∞ Control Synthesis – p.25/27

References

- Chen, Francis: Optimal Sampled-Data Control, Springer (1995)
- Zhou, Doyle, Glover: Robust and Optimal Control, Prentice Hall (1996)
- Skelton, Iwasaki, Grigoriadis: A Unified Algebraic Approach to Linear Control Design, Taylor & Francis (1998)
- Gahinet, Automatica, vol.32, pp.1007–1014 (1996)

Discrete-Time H_∞ Control Synthesis – p.27/27

Controller Construction Algorithm

Algorithm

■ **Step 0:** (X, Y) : solution to synthesis LMI

■ **Step 1:** Find D_K such that $\Delta(D_K) > 0$

■ **Step 2:** Find \tilde{B}_K and \tilde{C}_K such that

$$\begin{bmatrix} X & L_X(\tilde{C}_K) \\ L_X^\top(\tilde{C}_K) & \Delta(D_K) \end{bmatrix} > 0, \quad \begin{bmatrix} Y & L_Y(\tilde{B}_K) \\ L_Y^\top(\tilde{B}_K) & \Delta(D_K) \end{bmatrix} > 0$$

■ **Step 3:** Set $\tilde{A}_K = L_Y(\tilde{B}_K)\Delta^{-1}(D_K)L_X^\top(\tilde{C}_K)$

■ **Step 4:** Recover Z_K by

$$Z_K = \left[\begin{array}{cc} I & 0 \\ 0 & N^{-1} \end{array} \right] \left(\left[\begin{array}{cc} I & 0 \\ -YB_2 & I \end{array} \right] \left[\begin{array}{cc} D_K & \tilde{C}_K \\ \tilde{B}_K & \tilde{A}_K \end{array} \right] \left[\begin{array}{cc} I & -C_2X \\ 0 & I \end{array} \right] - \left[\begin{array}{cc} 0 & 0 \\ 0 & YAX \end{array} \right] \right) \left[\begin{array}{cc} I & 0 \\ 0 & M^{-\top} \end{array} \right]$$

♣ SVD-based methods available for Step 1 and 2

Discrete-Time H_∞ Control Synthesis – p.26/27